

TECHNICAL FIELD OF THE INVENTION

The present invention relates generally to communication systems, and more particularly to a system and method for providing optical amplification using input signals having launch powers that are a function of the noise figure of at least a portion of the system.

BACKGROUND

In designing a wavelength division multiplexed optical transmission link including multiple spans of fiber with optical amplifiers interposed between the spans, conventional design approaches assume that the noise figure for the system is spectrally flat and equal in magnitude to the worst case noise figure for the system. Designers calculate a desired signal to noise ratio (SNR) as a function of the number of spans in the system, and select a launch power for wavelength signals input to the system that ensures that all channels will achieve the desired SNR, even at wavelengths having the highest noise figure. Generally, designers apply the same launch power to all wavelength signals.

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SUMMARY OF EXAMPLE EMBODIMENTS

The present invention recognizes a need for a more efficient optical communications system and method of communicating signals.

5 In one embodiment, an optical amplifier comprises a gain medium operable to receive a plurality of signals each comprising a center wavelength and a noise figure associated with at least a portion of the amplifier and varying as a function of wavelength. At least two of the
10 plurality of signals comprise a launch power that is a function of a magnitude of the noise figure measured at or near the center wavelength of that signal.

In another embodiment an optical amplifier comprises an input operable to receive a plurality of signals each
15 comprising a center wavelength, wherein at least two of the plurality of signals comprise different launch powers. The amplifier further comprises a pump operable to generate a pump signal and a gain medium operable to receive the plurality of signals and the pump signal and
20 to facilitate amplification of at least some of the plurality of signals. The amplifier also comprises an output operable to communicate amplified versions of the plurality of signals. A signal to noise ratio measured at the output of the amplifier varies by no more than 2.5
25 decibels over a bandwidth of at least 40 nanometers for at least a majority of signals output from the amplifier.

In still another embodiment, an optical communication system comprises an input terminal comprising a plurality of optical transmitters each
30 operable to output one of a plurality of signals each comprising a center wavelength. The system further comprises a plurality of spans of optical medium coupled

to the input terminal and operable to facilitate communication of the plurality of signals and a plurality of in-line amplifiers each coupled to at least one of the plurality of spans of optical medium. At least some of the plurality of signals comprise a launch power that is a function of a noise figure associated with at least a portion of the system.

10 In a method embodiment, a method of communicating optical signals comprises communicating a plurality of signals each having a center wavelength to an optical link comprising a plurality of spans of fiber. The method further comprises amplifying the plurality of signals to at least partially compensate for losses in one or more of the plurality of spans of fiber. Signals output from the optical link experience a noise figure varying as a function of wavelength. At least two of the signals input to the optical link comprise a launch power that is a function of the noise figure measured at or near the center wavelength of that signal.

20 In another method embodiment, a method of communicating signals comprises adjusting launch powers of a plurality of signals input to an optical link based at least in part on a noise figure associated with at least a portion of the optical link. The method further comprises adjusting a pump power of an amplifier in the optical link to give a desired gain spectrum in light of the adjusted launch powers. The steps of adjusting the launch power and adjusting the pump power are repeated until a signal to noise ratio at an output from the optical link varies by no more than a threshold amount for at least a majority of signals output from the optical link.

Depending on the specific features implemented, particular embodiments may exhibit some, none, or all of the following technical advantages. One embodiment provides a mechanism for reducing the total launched signal power in an optical link. Reducing the launched signal power reduces the intensity of light on connectors and other components, increasing the reliability of the system. Additionally, reduced launched signal power allows for use of lower powered pumps in amplifiers within the system. Reducing the pump power required generally results in decreased system costs.

As an additional benefit this technique facilitates freedom in design of gain profiles in multiple stage amplifiers. Because signal launch power is selected to at least partially address the noise figure issue, gain profiles of the amplifiers can be selected with less regard to maintaining a particular noise figure shape or magnitude.

Although this technique applies to and benefits many amplifier types, at least the following additional advantages can be realized when applying this technique to systems using Raman amplification. For example, when implemented in a Raman amplification system, this technique results in reduced non-linear penalties, such as four-wave mixing and Brillouin effect, which tend to be less prevalent at lower signal powers.

Other technical advantages of the present invention will be readily apparent to one skilled in the art from the following figures, description and claims.

BRIEF DESCRIPTION OF THE DRAWINGS

For a more complete understanding of the present invention, and for further features and advantages thereof, reference is now made to the following description taken in conjunction with the accompanying drawings, in which:

FIGURE 1 is a block diagram showing an exemplary optical communication system implementing at least some aspects of the present invention;

FIGURES 2a-2b are graphs illustrating simulated gain and noise figure curves for an example amplifier design implementing approximately flat gain profiles in each amplifier stage;

FIGURES 3a-3b are graphs illustrating simulated gain and noise figure curves for another example amplifier design implementing approximately flat gain profiles in each amplifier stage;

FIGURES 4a-4c illustrate an exemplary embodiment of a multiple stage amplifier including at least two amplification stages, gain profiles associated with various amplification stages of the amplifier, and an overall gain profile for the amplifier, respectively;

FIGURES 5a-5c illustrate another exemplary embodiment of a multiple stage amplifier including at least two amplification stages, gain profiles associated with various amplification stages of the amplifier, and an overall gain profile for the amplifier, respectively;

FIGURES 6a-6c illustrate another exemplary embodiment of a multiple stage amplifier including at least three amplification stages, gain profiles associated with various amplification stages of the

amplifier, and an overall gain profile for the amplifier, respectively;

5 FIGURES 7a-7c illustrate another exemplary embodiment of a multiple stage amplifier including at least four amplification stages, gain profiles associated with various amplification stages of the amplifier, and an overall gain profile for the amplifier, respectively;

10 FIGURE 8 is a graph illustrating simulated results of one particular amplifier design implementing various combinations of gain profiles;

 FIGURE 9 is a graph illustrating simulated results of another amplifier design implementing various combinations of gain profiles; and

15 FIGURE 10 is a flow chart illustrating one example of a method of determining a launch power for a wavelength signal in a multiple span communication system.

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FOOTNOTES

DETAILED DESCRIPTION OF EXAMPLE EMBODIMENTS

FIGURE 1 is a block diagram showing an exemplary optical communication system 10 operable to facilitate communication of one or more multiple wavelength signals. System 10 can be configured to provide unidirectional or bi-directional communication of multiple wavelength signals. In this example, system 10 includes a transmitter bank 12 operable to generate a plurality of optical signals (or channels) 15a-15n, each comprising a center wavelength of light. In a particular embodiment, each optical signal 15 can comprise a center wavelength substantially different from the center wavelengths of other optical signals 15. As used throughout this document, the term "center wavelength" refers to a time averaged mean of the spectral distribution of an optical signal. The spectrum surrounding the center wavelength need not be symmetric about the center wavelength. Moreover, there is no requirement that the center wavelength represent a carrier wavelength.

Transmitters 12 could reside, for example, within a transponder capable of transmitting and receiving signals. In one embodiment, the "plurality of transmitters" of transmitter bank 12 comprises a plurality of independent pairs of optical sources and associated modulators. Alternatively, the "plurality of transmitters" could comprise one or more optical sources shared by a plurality of modulators. For example, transmitter bank 12 could comprise a continuum source transmitter including a modelocked source operable to generate a series of optical pulses and a continuum generator operable to receive a train of pulses from the modelocked source and to spectrally broaden the pulses to

form an approximate spectral continuum of signals. The continuum generator could operate, for example using soliton-effect compression or adiabatic soliton compression. A signal splitter receives the continuum and separates the continuum into individual signals each having a center wavelength. Modulators operate to encode information onto the signals received to produce signals 15 for transmission to optical communications medium 20. In some embodiments, transmitter bank 12 can also include a pulse rate multiplexer, such as a time division multiplexer, operable to multiplex pulses received from the mode locked source or the modulator to increase the bit rate of the system.

In the illustrated embodiment, system 10 also includes a combiner 14 operable to receive a plurality of optical signals 15a-15n and to combine those wavelength signals into a multiple wavelength signal 16. As one particular example, combiner 14 could comprise a wavelength division multiplexer (WDM). The terms wavelength division multiplexer and wavelength division demultiplexer as used herein may include equipment operable to process wavelength division multiplexed signals and/or dense wavelength division multiplexed signals.

System 10 communicates multiple wavelength optical signal 16 over an optical communication medium 20. Communication medium 20 can comprise a plurality of spans 20a-20n of fiber, each coupled to or comprising an optical amplifier. In some embodiments all or a portion of a span can serve as a distributed amplification stage. Fiber spans 20a-20n could comprise standard single mode fiber (SMF), dispersion-shifted fiber (DSF), non-zero

dispersion-shifted fiber (NZDSF), or another fiber type or combination of fiber types.

Two or more spans of communication medium 20 can collectively form an optical link. In the illustrated example, communication medium 20 includes a single optical link 25 comprising numerous spans 20a-20n. System 10 could include any number of additional links coupled to link 25. Although optical link 25 is shown to include one or more booster amplifiers 18 and preamplifiers 24, one or more of these amplifier types could be eliminated in other embodiments.

In this example, system 10 includes a booster amplifier 18 operable to receive and amplify wavelengths of signal 16 in preparation for transmission over a communication medium 20. Where communication system 10 includes a plurality of fiber spans 20a-20n, system 10 can also include one or more in-line amplifiers 22a-22n. In-line amplifiers 22 couple to one or more spans 20a-20n and operate to amplify signal 16 as it traverses communication medium 20. Optical communication system 10 can also include a preamplifier 24 operable to amplify signal 16 received from a final fiber span 20n.

Throughout this document, the term "amplifier" denotes a device or combination of devices operable to at least partially compensate for at least some of the losses incurred by signals while traversing all or a portion of optical link 25. Likewise, the term "amplification" refers to offsetting at least a portion of losses that would otherwise be incurred.

An amplifier may, or may not impart a net gain to a signal being amplified. Moreover, the term "gain" as used throughout this document, does not - unless

explicitly specified - require a net gain. In other words, it is not necessary that a signal experiencing "gain" or "amplification" in an amplifier stage experiences enough gain to overcome all losses in the amplifier stage. As a specific example, distributed Raman amplifiers stages typically do not experience a net gain because of the high losses in the transmission fiber that serves as a gain medium. Nevertheless, these devices are considered "amplifiers" because they offset at least a portion of the losses experienced in the transmission filter.

Amplifiers 18, 22, and 24 could each comprise, for example, a discrete Raman amplifier, a distributed Raman amplifier, a rare earth doped amplifier such as an erbium doped or thulium doped amplifier, a semiconductor amplifier or a combination of these or other amplifier types.

In some embodiments, multiple wavelength signal 16 carries optical signals 15a-15n having center wavelengths ranging across different communications bands (e.g., the short band (S-band), the conventional band (C-band), and/or the long band (L-band)). In those cases, amplifiers 18, 22, and 24 could each comprise a wide band amplifier, each operable to amplify all signal wavelengths received. Alternatively, one or more of those amplifiers could comprise a parallel combination of amplifier assemblies, each operable to amplify a portion of the wavelengths of multiple wavelength signal 16. In that case, system 10 could incorporate signal dividers and signal combiners surrounding the parallel combinations of amplifier assemblies to facilitate separation of the wavelength groups prior to

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power needed to achieve a desired SNR at the wavelength experiencing the highest noise figure. Applying a constant launch power to all wavelength signals, however, can be inefficient because the noise figure in implemented optical systems is not always constant.

For example, Raman interaction among the multiple wavelengths of signal 16 causes longer wavelength signals to capture energy from shorter wavelength signals, which tends to result in a larger noise figure at short wavelengths than at longer wavelengths. Moreover, at least in Raman amplifiers, phonon induced noise tends to occur at wavelengths near the wavelength of the pump signals for the amplifiers (typically near the shorter wavelengths being amplified), making the noise figure larger at or near those wavelengths. Therefore, designing an optical communication system assuming a constant noise figure and using a constant signal launch power can result in many channels (at least those at wavelengths experiencing less than the worst case noise figure) being launched at powers far in excess of that needed to achieve the desired SNR.

Other systems, rather than using constant launch powers, pre-emphasize launch powers to follow the loss of the fiber or to account for intra-band and/or interband Raman energy exchange. Those systems too, however, result in inefficiencies because all channels are designed with reference to a worst case noise figure. None of these other approaches implement launch powers determined as a function of noise figure.

The embodiment depicted in FIGURE 1 addresses this inefficiency by selecting at least some of the launch powers of optical signals 15a-15n to be different in

5 magnitude than others of optical signals 15a-15n. For example, the magnitude of the launch power of at least some of optical signals 15a-15n can be selected depending at least in part on the magnitude of the noise figure associated with at least a portion of system 10 at a wavelength corresponding to the center wavelengths of those optical signals 15.

10 In one particular embodiment, the launch power of each optical signal 15 input at a given position in system 10 can be separately selected depending at least in part on the magnitude of a noise figure corresponding to the associated wavelength and associated with all or a portion of system 10. Throughout this description, the phrase "launch power" refers to a signal's power at the
15 input to any portion of system 10 over which the noise figure will be measured for use in determining or modifying the signal's launch power.

20 Signal launch powers may be selected, for example, to achieve a desired SNR at an output location. The output location may comprise, for example, a receiver 28 at an end of optical link 25, or could comprise an access element coupled to link 25. Given the desired output SNR and knowing the noise figure associated with a particular wavelength, the launch signal power for that wavelength
25 can be determined.

30 It is not necessary that all optical signals 15 have launch powers independently selected with reference to the noise figure. Moreover, even for those wavelength signals having a launch power selected with reference to the noise figure at that wavelength, it is not necessary that each launch power be determined with reference to the same desired SNR. For example, in some cases it may

be desirable to select launch powers for one set of optical signals 15 to ensure obtaining a first SNR, while launch powers for another set of optical signals 15 are selected to ensure obtaining a second SNR, different than the first SNR.

This may be advantageous, for example, where some wavelengths are designated as long haul wavelengths, while others are designated for add/drop processing at access elements along link 25. The long haul signals, for example, may require a higher SNR than the signals traversing only a portion of link 25. The launch powers of the long haul signals, therefore, could be determined with reference to one SNR, while launch powers of the other signals are determined with reference to a different SNR. Any number of sets of wavelength signals could have their launch powers determined with respect to different SNRs, depending on the design criteria of the system.

As a particular example of determining the launch power for a signal, assume that an output SNR (in decibels) equals the difference between an input SNR (in decibels) and the noise figure (in decibels) at the output; or, $SNR_{out} = SNR_{in} - NF$. This relationship assumes that all of the signal to noise ratios are referred to the output of an ideal photo diode (e.g., 100% quantum efficiency) and is valid when the input light has the a shot noise limited signal to noise ratio ($SNR_{in} = SNR_{snl}$). The shot noise limited signal to noise ratio can be represented mathematically as:

$$SNR_{snl} = P_s + 10\log(\lambda/1 \text{ micron}) + 154.01\text{dB} - 10\log(BW/1\text{Hz})$$

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5. In that equation, P_s is the launch power of the signal in decibels above one milli-watt (dBm) (which equals $10\log(P_s)$ in milli-watts); λ is the wavelength of the signal in microns, and BW is the detection bandwidth used for the given wavelength in Hz. For a given bandwidth and a selected SNR_{out} , the launch power can be expressed as:

$$P_s = SNR_{out} + 10\log(BW/1Hz) - 154.01dB + NF - 10\log(\lambda/1 \text{ micron})$$

10 or, simplifying that expression:

$$P_s = C + NF - 10\log(\lambda/1 \text{ micron});$$

15 where C is a constant that depends on the bandwidth (BW) of the amplified signals and the desired output SNR. Thus, for a given bandwidth and a desired SNR, the launch power P_s of each optical signal 15 can be expressed as a function of the noise figure, that function having a
20 small dependence on the wavelength of the signal. For a desired SNR, the signal launch power can be determined based on the noise figure without reference to the ($\lambda/1$ micron) wavelength dependence, resulting in a small
25 variance in the SNR of the system (typically around 0.3 decibels over a bandwidth of 100 nanometers). Alternatively, the signal launch powers can be determined with reference to the noise figure and accounting for the wavelength dependence, resulting in a flat SNR.

30 Given one or more desired SNRs, the launch powers of optical signals 15 can be selectively determined through any of a variety of mechanisms. For example, a drive current applied to optical sources generating optical signals 15 could be adjusted. As another example, optical signals 15 could each be generated at a common

power, and applied to a variable attenuator operable to attenuate some wavelengths more than others to result in the desired distribution of launch powers.

5 A launch power that is selected with reference to the noise figure can be determined initially with respect to the noise figure, and/or may be adjusted from time to time with respect to the noise figure. For example, in some embodiments, signal launch powers are initially selected during system setup by comparing the signal powers with the noise figure and adjusting the signal launch powers accordingly. In other embodiments, signal launch powers are monitored continuously, periodically, or on a random basis during system operation and adjusted in power depending on the shape and magnitude of the noise figure at that time. These embodiments can help account for changes in the shape and/or magnitude of the noise figure due to changes in system characteristics over time, such as temperature variations or aging of components and addition/subtraction of channels being communicated.

20 The noise figure used to influence the magnitude of the launch powers of one or more of optical signals 15 can be determined at various locations within system 10. For example, the noise figure for the entire system 10 could be determined at receivers 28a-28n and used for determining launch powers for optical signals 15. Alternatively, the noise figure for a portion of system 10 could be measured at a location along link 25, such as an optical add/drop multiplexer, where one or more optical signals 15 are added or dropped from multiple wavelength signal 16.

5 This technique is not limited to controlling the launch power of signals generated at transmitters 12 associated with link 25. The technique could also be applied to signals 15 initially launched on other optical links and later combined with multiple wavelength signal 16 on optical link 25.

10 In some embodiments, system 10 can also comprise a management system 35 operable to track and/or manage various aspects of operation of system 10 and/or the components therein. For example, management system 35 could comprise hardware, software, firmware, or a combination thereof operable to selectively adjust the launch power to a optical signal 15 and/or a pump power applied to an amplifier in system 10. In some
15 embodiments, management system 35 can measure or receive a noise figure associated with all or a portion of system 10, determine launch powers for signals 15 based at least in part on the noise figure, and communicate control signals to adjust launch powers of one or more signals 15
20 accordingly.

25 Although the illustrated embodiment shows management system 35 directly coupled to each amplifier, transmitter, and receiver, management system could alternatively communicate with some or all of those devices via communication medium 20 using, for example, an optical service channel. Furthermore, management system 35 need not, in all embodiments, communicate with all amplifiers, transmitters, and receivers. Although management system 35 is depicted as a single entity
30 located remotely from amplifiers 18-24, all or a part of management system 35 could alternatively reside locally

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to one or more amplifiers, transmitters, and/or receivers in system 10.

Management system 35 could be used to aid in manually configuring system 10, or could operate to dynamically configure system 10 initially and/or as it continues to operate. Management system 35 could be used to initially set launch powers for optical signals 15 and/or to periodically, randomly, or on demand reconfigure launch powers of optical signals 15 to account for changing system parameters.

By setting launch powers of at least some of the optical signals 15 with reference to a noise figure associated with at least a portion of system 10, this embodiment provides an advantage of reducing the total launched signal power. Reducing the launched signal power reduces the intensity of light at the connectors and other components, increasing the reliability of system 10. Additionally, reduced launched signal power allows for use of lower powered pumps in amplifiers within system 10. Reducing the pump power required generally results in decreased system costs.

As an additional benefit this technique facilitates freedom in design of gain profiles in the multiple stage amplifiers. Because signal launch power is selected to at least partially address the noise figure issue, gain profiles of the amplifiers can be selected with less regard to maintaining a particular noise figure shape or level. For example, because the signal launch power settings can be adjusted to deal with the noise figure, a single stage amplifier with a flat gain profile or a multiple stage amplifier with flat gain profiles in each stage can be used without requiring excessive signal

launch powers. Alternatively, gain profiles in multiple stage amplifiers can be tailored to accomplish other functions, such as reducing the average noise figure of the amplifier or reducing the total pump power used by the amplifier. Various example multiple stage amplifier designs are discussed below with respect to FIGURES 4-7.

Although this technique applies to and benefits many amplifier types, at least the following additional advantages can be realized when applying this technique to systems using Raman amplification. For example, when implemented in a Raman amplification system, this technique results in reduced non-linear penalties, such as four-wave mixing and Brillouin effect, which tend to be less prevalent at lower signal powers. Furthermore, at least in Raman amplifiers, the noise figure tends to be larger at shorter wavelengths. Using the above-described technique, shorter wavelengths will often have larger launch powers to compensate for the larger noise figure. The increased power in shorter optical signals promotes signal-signal interaction and transfer of some of the launch power of shorter optical signals to longer optical signals, further reducing the pump power necessary to achieve a desired gain.

FIGURE 2a is a graph illustrating simulated gain and noise figure curves for a two stage Raman amplifier design implementing approximately flat gain profiles in each amplifier stage. In particular, this simulation assumes a two stage Raman amplifier having a first stage comprising a distributed Raman amplification stage utilizing approximately eighty kilometers of SMF-28 fiber, and a second stage comprising a discrete Raman amplification stage utilizing a length of DK-80

dispersion compensating fiber. The gain profiles for each amplifier stage are substantially flat, simplifying the amplifier design. Table 1 below shows pump wavelength locations and powers for each stage.

5 TABLE 1

Flat Profile Two Stage Amplifier Applying 0.416 mW / Channel		Flat Profile Two Stage Amplifier Applying Varying Signal Power	
Pump λ	Power (W)	Pump λ	Power (W)
80 km SMF-28		80 km SMF-28	
1396 nm	.56	1396 nm	.56
1416 nm	.56	1416 nm	.56
1427 nm	.56	1427 nm	.56
1455 nm	.25	1455 nm	.25
1472 nm	.1	1472 nm	.1
1505 nm	.085	1505 nm	.085
DK-80		DK-80	
1405 nm	.47	1405 nm	.47
1418 nm	.53	1418 nm	.53
1445 nm	.31	1445 nm	.31
1476 nm	.085	1476 nm	.085
1509.5 nm	.025	1509.5 nm	.025
Total Pump Power: 3.535 W		Total Pump Power: 3.535 W	

This example simulates results for the amplifier in two configurations. In both cases, the total launched signal power among 250 optical signals 15 was 104 milli-watts. In the first configuration, each of optical signals 15 was launched at 0.416 milli-watts. In the second configuration, the total launched power was distributed among the optical signals 15 with reference to the noise figure of the amplifier to achieve an SNR of approximately 33.2 decibels, resulting in higher launched signal powers at shorter wavelengths where the noise figure was larger. In particular, launch powers of optical signals 15 were determined by applying the following equation:

$$P_s = -25.6\text{dBm} + \text{NF (for each signal wavelength)}.$$

5 Note that this equation does not consider the
10 $10\log(1/\text{micron})$ wavelength dependence. As a result,
there will be a slight variation in the SNR as a function
of wavelength.

10 Line 17 in FIGURE 2a shows the overall gain curve
and line 19 shows the noise figure for the first
embodiment (constant launch power). Line 31 in FIGURE 2a
shows the overall gain curve and line 33 shows the noise
figure for the second embodiment (variable launch power).
As shown in this figure, varying the signal power as a
function of the amplifier noise figure does not result in
15 any significant penalty in terms of peak noise figure or
flatness of gain curve. In fact, in this embodiment,
varying the launched power as a function of the noise
figure results in a lower peak noise figure and an
increased gain level. The lower peak noise figure is
20 likely attributable to the lower launched power, which
reduces noise caused by signal-signal interactions in the
amplifier.

25 FIGURE 2b is a graph showing the SNR resulting from
the simulations shown in FIGURE 2a and assuming a
receiver detection bandwidth of 5 gigahertz. In
particular, line 37 shows the SNR for the constant launch
power embodiment, while line 39 shows the SNR for the
variable launch power embodiment. As shown in this
figure, the variable launch power embodiment results in
30 an approximately flat SNR across the amplified bandwidth,
varying by about 0.3 decibels from 1515 nanometers to
1625 nanometers. This logically follows from the fact

that the launch powers of optical signals 15 were selected based on the noise figure at each wavelength.

5 The slight variance in SNR 39 (the varying launch power embodiment) results from the slight wavelength dependence of the SNR, which was not considered in determining launch powers in this example. In another embodiment, signal powers could be selected based on a combination of the variance in the noise figure and the center wavelength of the signal 15, to result in a completely flat SNR.

10 The constant launch power embodiment shows a significantly varying SNR 37 over the bandwidth of amplified wavelengths, varying by over five decibels. This results because the signal launch powers remain constant while the noise figure varies as a function of wavelength. Because the launch power remained constant in that embodiment, SNR 37 is lowest where the noise figure is the highest.

15 At those wavelengths, SNR 37 is significantly below SNR 39. For example, at approximately 1522 nanometers, the SNR 39 for the variable launch power embodiment is more than three decibels higher than the SNR 37 for the constant launch power embodiment. If one were to use a constant launch power embodiment and require an SNR equal to that of the variable launch power embodiment across the entire amplified bandwidth, it would be necessary to increase the launch power of all optical signals 15 by more than three decibels in the constant launch power embodiment. This results in inefficiency because launch powers would be increased unnecessarily where the noise figure is low. The variable launch power embodiment,

therefore, can result in efficiencies over a constant launch power approach.

FIGURE 3a is a graph illustrating simulated gain and noise figure curves for another embodiment of a two stage Raman amplifier design implementing approximately flat gain profiles in each amplifier stage. In particular, this simulation assumes a two stage Raman amplifier having a first stage comprising a distributed Raman amplification stage utilizing approximately eighty kilometers of non-zero dispersion shifted fiber (NZDSF), and a second stage comprising a discrete Raman amplification stage utilizing a length of DK-80 dispersion compensating fiber. The gain profiles for each amplifier stage are substantially flat. Table 2 below shows pump wavelength locations and powers for each stage.

TABLE 2

Flat Profile Two Stage Amplifier Applying 0.430 mW / Channel Pump λ Power (W)	Flat Profile Two Stage Amplifier Applying Varying Signal Power Pump λ Power (W)
80km NZDSF	80km NZDSF
1396 nm .343	1396 nm .343
1416 nm .343	1416 nm .343
1427 nm .343	1427 nm .343
1455 nm .153	1455 nm .153
1472 nm .0612	1472 nm .0612
1505 nm .052	1505 nm .052
DK-80	DK-80
1405 nm .47	1405 nm .47
1418 nm .55	1418 nm .55
1445 nm .33	1445 nm .33
1476 nm .083	1476 nm .083
1509.5 nm .023	1509.5 nm .023
Total Pump Power: 2.7512 W	Total Pump Power: 2.7512 W

This example simulates results for the amplifier in two configurations. In both cases, the total launched signal power among 250 optical signals 15 was approximately 107 milli-watts. In the first configuration, each of optical signals 15 was launched at 0.430 milli-watts. In the second configuration, the total launched power was distributed among the optical signals 15 with reference to the noise figure of the amplifier to achieve an SNR of approximately 33.2 decibels, resulting in higher launched signal powers at shorter wavelengths where the noise figure was larger. In particular, launch powers of optical signals 15 were determined by applying the following equation:

$$P_s = -25.6 \text{ dBm} + \text{NF}.$$

Line 117 in FIGURE 3a shows the overall gain curve and line 119 shows the noise figure for the first embodiment (constant launch power). Line 131 in FIGURE 3a shows the overall gain curve and line 133 shows the noise figure for the second embodiment (variable launch power). As shown in this figure, varying the signal power as a function of the amplifier noise figure does not result in any significant penalty in terms of peak noise figure or flatness of gain curve, but rather results in a lower peak noise figure and an increased gain level.

FIGURE 3b is a graph showing the SNR resulting from the simulations shown in FIGURE 3a and assuming a receiver detection bandwidth of 5 gigahertz. In particular, line 137 shows the SNR for the constant launch power embodiment, while line 139 shows the SNR for

the variable launch power embodiment. Again, the variable launch power embodiment results in an approximately flat SNR 139 across the amplified bandwidth, in this case varying by approximately 0.3
5 decibels from 1515 nanometers to 1625 nanometers. Again, in this example, signal launch powers were selected without reference to the wavelength dependence of the SNR resulting in a slight variance in the SNR. Accounting for this wavelength dependence can result in a completely
10 flat SNR.

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15 The constant launch power embodiment shows a significantly varying SNR 137 over the bandwidth of amplified wavelengths, varying by more than six decibels. Because the launch power remained constant in that embodiment, SNR 137 is lowest where the noise figure is the highest. At those wavelengths, SNR 137 is significantly below SNR 139. For example, at approximately 1522 nanometers, the SNR 139 for the variable launch power embodiment is nearly four decibels
20 higher than the SNR 137 for the constant launch power embodiment. To use a constant launch power embodiment and ensure an SNR equal to that of the variable launch power embodiment across the entire amplified bandwidth, it would be necessary to increase the launch power of all
25 optical signals 15 by nearly four decibels, resulting in significant additional launch power.

30 Although FIGURES 2 and 3 were described with respect to a Raman amplification system, similar results can be obtained using any amplifier type. The examples described herein with respect to Raman amplification systems are presented for illustrative purposes only.

As discussed above, varying signal launch powers with reference to a noise figure of all or a portion of the system provides an advantage of allowing the use of flat gain profiles in the amplifiers. This facilitates simplification of amplifier design and can result in more inexpensive amplifiers. At the same time, this approach ensures a desired SNR without requiring excessive launch powers across all amplified wavelengths.

As further indicated above, addressing the noise figure issue through varying launch powers allows freedom of design in amplifier gain profiles to address various design concerns. Using launch powers that vary by wavelength, multiple stage amplifiers can implement gain profiles in each amplification stage that are tailored to bring about particular desirable results. For example, gain profiles can be selected to reduce the average noise figure of the amplifier, or to reduce the pump power required to provide a particular gain.

FIGURES 4a-4c illustrate an exemplary embodiment of a multiple stage amplifier 100 including gain profiles 30 and 40 associated with various amplification stages and an overall gain profile 50 for the amplifier. The embodiment shown in FIGURE 4a provides an example of a multiple stage amplifier 100 receiving variable launch power signals and implementing non-flat gain profiles in at least some stages to reduce the peak noise figure of the amplifier. By reducing the peak noise figure, amplifier 100 facilitates reducing the launch power needed to achieve a given SNR even where the signal launch powers are held constant across the amplification bandwidth. Reducing the signal launch power reduces stress on system components and reduces pump powers

needed to generate the correspondingly lower powered output signals.

5 While the examples described with respect to FIGURES 2 and 3 comprise multiple stage amplifiers having approximately flat gain profiles in each stage, the amplifiers depicted in FIGURE 4 comprises multiple amplification stages having varying gain profiles, which are approximately complimentary to one another.

10 Throughout this description, the phrase "approximately complementary" refers to a situation where, at least in general, wavelength signals that are highly amplified in the first stage are less amplified in the second stage, and wavelength signals that are highly amplified in the second stage are less amplified in the first stage. Two gain profiles said to be "approximately complementary" need not have equal and opposite slopes. Moreover, equal amplification of any particular wavelengths in both gain profiles does preclude those gain profiles from being "approximately complementary."

20 Approximately complementary gain profiles may have one or more slopes associated with each gain profile. For example, approximately complementary gain profiles could comprise a "W" shaped profile followed by an "M" shaped profile, or an "M" shaped profile followed by a "W" shaped profile. Furthermore, the approximately complementary gain profiles may become approximately complementary only after traversing all or a portion of the transmission medium. In those cases, the gain profiles launched at the beginning of the amplifier stage may not be approximately complementary, but may become approximately complementary after signals traverse all or a portion of the transmission medium.

30

While best results are obtained by applying approximately complimentary gain profiles to all or nearly all of the same signal wavelengths, some portion of wavelengths can be omitted from one gain profile and included in the other gain profile without departing from the scope of this invention.

Conventional designs of multi-stage amplifiers have experienced difficulties processing bandwidths in excess of 80 nanometers while maintaining approximately flat gain profiles and acceptable noise figures. For example, in Raman amplifiers, a major culprit in noise figures is the phonon-stimulated optical noise created when wavelength signals being amplified reside spectrally close to pump wavelengths used for amplification. The embodiment shown in FIGURE 4a reduces adverse effect of this noise by enhancing the Raman amplification of signal wavelengths near the pump wavelengths to overcome the effects of the noise, and applying an approximately complementary gain profile in another stage to result in an approximately flat overall gain profile with a reduced noise figure.

In this example, amplifier 100 comprises a two-stage amplifier having a first stage 112 and a second stage 114 cascaded with first stage 112. There is no limit to a particular number of amplifier stages. For example, additional amplification stages could be cascaded onto second stage 114. Moreover, although the illustrated embodiment shows second stage 114 cascaded directly to first stage 112, additional amplification stages could reside between first stage 112 and second stage 114 without departing from the scope of the invention.

Amplifier 100 could comprise a distributed Raman amplifier, a discrete Raman amplifier, a hybrid Raman amplifier which comprises both discrete and distributed stages, a rare earth doped amplifier, a semiconductor amplifier, or another amplifier type or combination of amplifier types. Each stage 112, 114 of amplifier 100 includes an input operable to receive a multiple wavelength optical input signal 116. As a particular example, signal 116 could include wavelengths ranging over one hundred nanometers.

Each stage 112, 114 also includes a gain medium 120, 121. Depending on the type of amplifier being implemented, media 120, 121 may comprise, for example a gain fiber or a transmission fiber. In a particular embodiment, media 120, 121 may comprise dispersion compensating fibers.

Each stage 112, 114 further includes one or more wavelength pumps 122. Pumps 122 generate pump light 124 at specified wavelengths, which are pumped into distributed gain media 120, 121. Pumps 122 may comprise, for example, one or more laser diodes. Although the illustrated embodiment shows the use of counter propagating pumps, under at least some circumstances using a relatively quiet pump, co-propagating pumps could also be used without departing from the scope of the invention.

In one particular embodiment, pump wavelengths 124 can be selected so that the longest wavelength pump signal 124 has a wavelength that is shorter than the shortest wavelength of signal 116. As one specific example, the longest wavelength of pump light 124 could be selected to be, for example, at least ten nanometers

shorter than the shortest wavelength of signal 116. In this manner, amplifier 100 can help to avoid phonon stimulated noise that otherwise occurs when pump wavelengths interact with wavelengths of the amplified signal.

Couplers 118b and 118c couple pump wavelengths 124a and 124b to gain distributed media 120 and 125, respectively. Couplers 118 could comprise, for example, wavelength division multiplexers or optical couplers. A lossy element 126 can optionally reside between amplifier stages 112 and 114. Lossy element 126 could comprise, for example, an isolator, an optical add/drop multiplexer, or a gain equalizer.

The number of pump wavelengths 124, their launch powers, their spectral and spatial positions with respect to other pump wavelengths and other wavelength signals, and the bandwidth and power level of the signal being amplified can all contribute to the shape of the gain profile for the respective amplifier stage.

FIGURE 4b shows exemplary gain profiles for first stage 112 and second stage 114. Gain profile 30 shows the overall gain of first stage 112 of amplifier 100 for a bandwidth ranging from the shortest wavelength of signal 116 (λ_{sh}) to the longest wavelength of signal 116 (λ_{lg}). Gain profile 40 shows the overall gain of second stage 112 of amplifier 100 for a bandwidth ranging from the shortest wavelength of signal 116 (λ_{sh}) to the longest wavelength of signal 116 (λ_{lg}). Each of gain profiles 30 and 40 reflects the effects of the other gain profile acting upon it.

In this example, gain profile 30 of first stage 112 has primarily a downward slope, where a majority of the

shorter signal wavelengths 116 are amplified more than a majority of the longer signal wavelengths 116. Conversely, gain profile 40 of second stage 114 is approximately complimentary to gain profile 30 of first stage 112. Gain profile 40 exhibits primarily an upward slope where a majority of the longer signal wavelengths 116 are amplified more than a majority of the shorter signal wavelengths 116.

Although gain profiles 30 and 40 are for simplicity depicted as each having substantially one slope, the slope of each gain profile may change numerous times. Moreover, it is not necessary that the entire slope of gain profile 30 be negative, or that the entire slope of gain profile 40 be positive. Each profile may exhibit any number of peaks and valleys over the amplified bandwidth.

Gain profile 50 (shown in dotted lines in FIGURE 4c) represents an exemplary overall gain profile of amplifier 100 resulting from the application of gain profiles 30 and 40 to signal 116. Overall gain profile 50 is approximately flat over at least substantially all of the bandwidth of wavelengths within signal 116.

This particular example provides a significant advantage in reducing the peak noise figure associated with the amplifier using complementary gain profiles. The complementary gain profiles reduce the peak noise figure by amplifying signals closest to the pump wavelengths at higher levels the signals at wavelengths far from the pump wavelengths. In addition, the noise figure is reduced by amplifying longer wavelength signals in a later amplifier stage. Moreover, implementing varying launch powers reduces the total launched signal

power, which, in Raman amplifiers, reduces noise generated from the signal-signal interactions. In a discrete amplifier embodiment, using this type of configuration, the noise figure of amplifier 100 in the small signal limit can be reduced to less than eight decibels, in some cases 7 decibels, even where the bandwidth of signal 16 exceeds 100 nanometers.

Complementary gain profiles can also be used to reduce the pump power requirements for a given amplifier. When this technique is combined with a technique of varying signal launch powers with reference to the noise figure, a high efficiency system can result, which uses relatively lower total pump power and relatively lower total signal launch power.

FIGURES 5a-5c illustrate a high pump efficiency embodiment of a multiple stage amplifier 110 including exemplary gain profiles 130 and 140 associated with various amplification stages and an overall gain profile 150 for the amplifier. Amplifier 110 shown in FIGURE 5a is similar in structure and function to amplifier 100 shown in FIGURE 4a. Like amplifier 100 shown in FIGURE 4a, amplifier 110 of FIGURE 5a includes a first amplification stage 112 and a second amplification stage 114. Each of stages 112 and 114 includes a gain medium 120, 121, respectively, which is operable to receive multiple wavelength input signal 116 and pump wavelengths 124a and 124b, respectively. Each amplifier stage 112 and 114 operates to amplify wavelengths of signal 116 according to gain profiles 130 and 140 as shown. In this example, at least first stage 112 comprises a Raman amplification stage. Second stage 114 could comprise a

Raman amplification stage, or another type of amplification stage.

5 The example shown in FIGURE 5 differs from the example shown in FIGURE 4 in that gain profile 130 (shown in FIGURE 5b) of first stage 112 exhibits primarily an upward slope where a majority of longer wavelengths of signal 116 are amplified more than the majority of shorter wavelengths of signal 116. Conversely, gain profile 140 of second stage 114 comprises an approximately complementary gain profile to first gain profile 130 of first stage 112. Profile 140 applies a higher gain to a majority of shorter wavelengths than the gain applied to the majority of longer signal wavelengths 116. In addition, in this embodiment, the power of pumps 122a driving first gain profile 130 can be reduced.

10 The Raman scattering effect transfers energy from shorter wavelength signals to longer wavelength signals. This embodiment leverages that fact to allow the longer pump wavelengths of Raman first stage 112 to accept energy from the shorter pump wavelengths of second stage. In a particular embodiment, amplifier 110 may include a shunt 160 between second gain medium 121 and first gain medium 120 to facilitate the longer pump wavelengths of first stage 112 accepting power from the shorter pump wavelengths of second stage 114. The combined effects of first stage 112 and second stage 114 result in an overall gain profile 150 (FIGURE 5c) of the amplifier that remains approximately flat.

20 This embodiment provides significant advantages in terms of efficiency by allowing the use of fewer wavelength pumps 122a in the first stage 112, and/or also by allowing each pump 122a to operate at a lower launch

power. By selecting signal launch powers with reference to the noise figure of the amplifier, this embodiment enjoys the further efficiency of reduced overall launched signal power.

5 The embodiment shown in FIGURE 5a can also provide improvements for the noise figure of the amplifier. For example, phonon stimulated noise is created in Raman amplifiers where wavelengths being amplified spectrally reside close to a wavelength of pump signals 124. By
10 spectrally separating pump wavelengths 124 from signal wavelengths 116, phonon stimulated noise can be reduced.

In a particular embodiment, pump wavelengths 124 are selected to have wavelengths at least 10 nanometers shorter than the shortest wavelength in signal 116 being
15 amplified. Moreover, in a particular embodiment, second stage 114, where a majority of the gain to short wavelengths of signal 116 is applied, comprises the last stage of amplifier 110.

Although the embodiments shown in FIGURES 4-5 show
20 two complementary amplification stages, additional complementary amplification stages could also be implemented.

FIGURE 6a is a block diagram of a three stage
25 amplifier 200 including gain profiles 230, 240, and 245 associated with various amplification stages, and an overall gain profile 250 for the amplifier. Amplifier 200 is similar in structure and function to amplifier 100 of FIGURE 4 but includes three cascaded amplification stages 212, 214, and 215. Each of amplifier stages
30 212-215 includes a gain medium 220, 221, 223, respectively, which operates to receive multiple wavelength signal 216 and pump wavelengths 224a-224c from

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pumps 222a-222c. At least second amplifier stage 214 comprises a Raman amplification stage. Each amplifier stage includes an optical coupler operable to introduce pump wavelengths 224 to the respective gain media. In some embodiments, lossy elements 226 may reside between one or more amplification stages 212-215. Lossy elements 226 may comprise, for example, optical add/drop multiplexers, isolators, and/or gain equalizers.

In this particular example, first stage 212 and second stage 214 operate in a similar manner to amplifier 100 shown in FIGURE 4a. In particular, first stage 212 applies a gain profile 230 that amplifies a majority of shorter signal wavelengths 216 more than it amplifies a majority of longer signal wavelengths 216. Second stage 214, conversely, applies and approximately complimentary gain profile 240 to signal 216, where the majority of longer wavelengths of signal 216 are amplified more than a majority of shorter wavelengths of signal 216.

The combination of second stage 214 and third stage 215, on the other hand, operates similarly to amplifier 110 shown in FIGURE 5a. While second stage 214 applies gain profile 240 amplifying a majority of longer signal wavelengths 216 more than a majority of shorter signal wavelengths 216, third stage 215 applies to gain profile 245, which amplifies a majority of shorter signal wavelengths 216 more than a majority of longer signal wavelengths 216. The composite gain profile 250 (shown in FIGURE 6c) resulting from the combination of amplifications in first, second, and third amplifier stages of amplifier 200 results in an approximately flat overall gain profile for the amplifier.

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5 This particular example reaps the efficiency
benefits discussed with respect to FIGURE 5, and permits
use of the noise figure reduction techniques discussed
with respect to FIGURES 4 and 5. For example, efficiency
10 advantages are realized by allowing longer pump
wavelengths 224b of second stage 214 to accept power from
high powered shorter pump wavelengths 224c of third
amplification stage 215. This results from the Raman
effect wherein longer wavelength signals accept energy
15 from shorter wavelength signals. As a result, second
stage 214 can be operated with fewer wavelength pumps
than what otherwise be required, and also with lower pump
launch powers.

20 In terms of improvements in noise figure, the gain
profiles of first stage 212 compared to second stage 214
result in high amplification of shorter wavelengths of
signal 216 to overcome phonon stimulated noise associated
with interaction of those signals with the longer pump
wavelengths 224a. In addition, providing a significant
25 amount of amplification to shorter wavelengths of signal
216 in the last stage 215 of amplifier 220 helps to
minimize the noise figure associated with amplifier 200.

Moreover, applying varied signal launch powers
depending at least in part on the noise figure of the
30 amplifier results in reducing the total signal launch
power, further increasing the efficiency of the system.

FIGURES 7a-7c illustrate a four stage amplifier,
gain profiles associated with various stages of the
amplifier, and a composite gain of the amplifier
35 respectively. In this example, amplifier 300 includes
four amplification stages 312, 314, 315, and 317. At

least third stage 315 comprises a Raman amplification stage.

As shown in FIGURE 7b, first stage 312 applies a gain profile 330 where a majority of shorter signal wavelengths are amplified more than a majority of longer signal wavelengths, and second stage 314 applies an approximately complimentary gain profile 335 where a majority of longer signal wavelengths are amplified more than a majority of shorter signal wavelengths. In this particular embodiment, the composite gain from first stage 312 and second stage 314 results in an approximately flat overall gain profile at the output of second stage 314.

Because the composite gain curve for the amplifier is approximately flat, this design advantageously facilitates addition and subtraction of particular wavelengths of signal 316 without the need for further manipulation of the gain. In addition, first and second gain stages 312 and 314 provide a low noise figure, reducing the effects of phonon stimulated noise in shorter wavelength signals closest to the pump wavelengths.

Particular wavelengths of signal 316 may be substituted with other wavelengths at access element 326b. After processing by access element 326b, signal 316 continues to third amplification stage 315, where gain profile 340 is applied as shown in FIGURE 7b. Signal 316 is then communicated to fourth stage 317 where gain profile 345 is applied to wavelengths of signal 316. Amplified signal 316 is then output at output port 365.

Third and fourth amplification stages of amplifier 300 are similar in structure and function to amplifier

110 described with respect to FIGURE 5. Through the use of this configuration, third and fourth amplifier stages 315 and 317 provide increased efficiency in operation. In particular, pump 322 can operate with fewer pump signals and/or lower pump power as a result of the Raman scattering effect which allows longer pump wavelengths 324c of Raman third stage 315 to accept power from shorter pump wavelengths 324d of fourth amplification stage 317. Moreover, third and fourth amplification stages 315 and 317 assist in maintaining a low noise figure by applying a significant amount of the gain to the shortest wavelengths of signal 316 at the last amplifier stage 317.

As in other embodiments, applying varied signal launch powers depending at least in part on the noise figure of the amplifier results in reducing the total signal launch power, further increasing the efficiency of the system.

Amplifiers depicted in FIGURES 4-7 can comprises wide band amplifiers operable to receive and amplify a wide bandwidth of multiple wavelength signal 16. In particular embodiments, the amplifiers can process over 80 nanometers of bandwidth, and in some cases over 100 nanometers of bandwidth while maintaining an approximately flat overall gain profile over the bandwidth of amplified signal wavelengths 16.

Throughout this document, the term "approximately flat overall gain profile" describes a condition where the maximum signal gain at the output of the amplifier differs from the minimum signal gain at the output of the amplifier by an no more than amount suitable for use in telecommunication systems over an operational bandwidth

of information carrying channels. Deviation of the maximum and minimum signal gain over one or two of several channels is not intended to be outside of the scope of an approximately flat overall gain profile. The deviation between minimum and maximum signal gains may comprise, for example five decibels prior to application of any gain flattening filters over an operational bandwidth of, for example, 40 nanometers or more. Particular embodiments of the invention may achieve gain flatness of approximately three decibels or less prior to application of any gain flattening filters over an operational bandwidth.

Implementing launch signal powers that vary by wavelength according to a noise figure associated with at least part of the system in combination with various amplifier gain profiles provides significant freedom in designing wide bandwidth amplifiers. As discussed above, varying signal launch powers can be combined with flat gain profile amplifiers to provide a simple amplifier design, which ensures a desired SNR, while minimizing the total launched signal power. In other embodiments, signal launch powers that vary by wavelength with the noise figure can be combined with more complex amplifier designs having sloped gain profiles in multiple amplifier stages. Combining varying signal power with these techniques can result in lower noise figures, or can provide a more efficient amplifier design, reducing both the launched signal power and the pump power required by the amplifier.

FIGURE 8 is a graph illustrating simulated results of one particular amplifier design implementing various combinations of gain profiles. This example assumes a

two stage Raman amplifier comprising a distributed Raman amplification stage followed by a discrete Raman amplification stage. The distributed Raman amplification stage implements approximately eighty kilometers of SMF-28 fiber, while the discrete Raman amplification stage implements a length of DK-80 dispersion compensating fiber.

In all cases, the launched signal power of each wavelength signal is varied depending on the noise figure of the amplifier to achieve an SNR equivalent to a system with one milli-watt per channel of signal power and a flat noise figure of 25.6 decibels (approximately 33.2 decibels with a 5 gigahertz detection bandwidth at 1520 nanometers). The varied launched signal power is applied to the amplifier in three configurations. The first configuration is one where the gain profiles of the amplification stages are approximately flat. This will be referred to as the "flat profile configuration."

A second configuration implements approximately complementary gain profiles in the first and second amplification stages, similar to those shown in FIGURE 4b. This will be referred to as the "low noise configuration."

A third configuration implements approximately complementary gain profiles in the first and second amplification stages similar to those shown in FIGURE 5b. This will be referred to as the "high pump efficiency configuration."

Table 3 below shows the pump wavelengths and powers applied in each amplification stage.

TABLE 3

<u>LOW NOISE</u>		<u>FLAT PROFILE</u>		<u>HIGH PUMP EFF.</u>	
Applying 147 mW		Applying 104 mW		Applying 128 mW	
Total Signal		Total Signal		Total Signal	
Power Varying By		Power Varying By		Power Varying By	
Wavelength		Wavelength		Wavelength	
Pump λ	Power	Pump λ	Power	Pump λ	Power
80 km SMF-28		80 km SMF-28		80 km SMF-28	
1396 nm	.56	1396 nm	.56	1396 nm	.56
1416 nm	.56	1416 nm	.56	1427 nm	.56
1427 nm	.56	1427 nm	.56	1455 nm	.25
1455 nm	.2	1455 nm	.25	1472 nm	.15
1472 nm	.08	1472 nm	.1	1505 nm	.25
1505 nm	.023	1505 nm	.085		
DK-80		DK-80		DK-80	
1405 nm	.56	1405 nm	.47	1396 nm	.56
1418 nm	.56	1418 nm	.53	1416 nm	.56
1445 nm	.4	1445 nm	.31	1427 nm	.22
1476 nm	.16	1476 nm	.085	1445 nm	.22
1509.5 nm	.063	1509.5 nm	.025	1476 nm	.04
				1509.5 nm	.0107
Total Pump		Total Pump		Total Pump	
Power: 3.726 W		Power: 3.535 W		Power: 3.3807 W	

Line 310 represents the gain curve for the flat profile configuration, while line 312 represents the noise figure for the flat profile configuration. Line 320 represents the gain curve for the low noise configuration, while line 322 represents the noise figure for that configuration. Line 330 represents the gain curve for the high pump efficiency configuration, while line 332 represents the noise figure for that configuration.

As can be appreciated from FIGURE 8, the low noise configuration enjoys the lowest peak noise figure. This is due at least in part to the complementary gain profiles used, which provides higher amplification to the wavelength signals closest in wavelength to the pump wavelengths, thus overcoming phonon induced noise. The low noise configuration, however in this example,

utilizes the most total pump power and the most total launched signal power.

5 The flat profile configuration experiences a slightly higher peak noise figure, but enjoys the lowest average noise figure. In addition, the flat profile configuration utilizes less pump power than the low noise configuration and uses the least launched signal power of any of these examples. The reduced launched signal power reduces the signal-signal interactions, which results in a lower average noise figure for the amplifier. 10 Moreover, the noise figure, and thus signal power, being highest at shorter wavelengths and lowest at longer wavelengths also results in a lower average noise figure.

15 The high pump efficiency configuration utilizes slightly more launched signal power than the flat profile configuration, but uses the least total pump power of all of the examples. Decreased pump power facilitates use of lower powered, less expensive pumps, or fewer pumps.

20 In any case, all of these designs can result in a substantially flat overall gain curve for the amplifier across a bandwidth of over eighty nanometers, in some embodiments over 100 nanometers. At the same time, the noise figure associated with the amplifier either is maintained at an acceptable level for all wavelengths, or 25 is not a problem because launched signal powers are selected to provide a desired SNR given the noise figure at each wavelength. This results in effective system performance, while minimizing the required signal launch power.

30 FIGURE 9 is a graph illustrating simulated results of a similar amplifier design as that shown in FIGURE 8. This example assumes a two stage Raman amplifier

comprising a distributed Raman amplification stage followed by a discrete Raman amplification stage. The distributed Raman amplification stage implements approximately eighty kilometers of non-zero dispersion shifted fiber (NZDSF), while the discrete Raman amplification stage implements a length of DK-80 dispersion compensating fiber. Table 4 shows pump powers used in the example from FIGURE 9:

TABLE 4

LOW NOISE		FLAT PROFILE		HIGH PUMP EFF.	
Applying 151 mW		Applying 107 mW		Applying 131 mW	
Total Signal		Total Signal		Total Signal	
Power Varying By		Power Varying By		Power Varying By	
Wavelength		Wavelength		Wavelength	
Pump λ	Power	Pump λ	Power	Pump λ	Power
80km NZDSF		80km NZDSF		80km NZDSF	
1396 nm	.38	1396 nm	.343	1396 nm	.343
1416 nm	.38	1416 nm	.343	1427 nm	.343
1427 nm	.35	1427 nm	.343	1455 nm	.153
1455 nm	.1	1455 nm	.153	1472 nm	.092
1472 nm	.05	1472 nm	.0612	1505 nm	.153
1505 nm	.0085	1505 nm	.052	DK-80	
DK-80		DK-80		1396 nm	.56
1405 nm	.38	1405 nm	.47	1416 nm	.56
1418 nm	.45	1418 nm	.55	1427 nm	.20
1445 nm	.53	1445 nm	.33	1445 nm	.23
1476 nm	.2	1476 nm	.083	1476 nm	.035
1509.5 nm	.09	1509.5 nm	.023	1509.5 nm	.0095
Total Pump		Total Pump		Total Pump	
Power: 2.9185 W		Power: 2.7512 W		Power: 2.6785 W	

FIGURE 10 is a flow chart illustrating one example of a method 500 of determining a launch power for a wavelength signal in a multiple span communication system. For ease of description, method 500 will be described with reference to communication system 10 shown in FIGURE 1. Method 500 could, however, apply to other optical communication systems, subsystems, or amplifiers.

The noise figure of an amplifier is affected by the power of the signals input to the amplifier and by the power of the pump signals supplied by the amplifier. Changing the launch power of input signals with reference to the noise figure can, therefore, change the noise figure making it desirable to further change the launch power of the input signals and/or pump signals.

With this in mind, method 500 begins at step 510 by adjusting the launch power of at least some of the signals 15 as a function of the noise figure of at least a portion of optical link 25. This could involve, for example, adjusting the launch power of some or all of signals 15 according to the following equation:

$$P_s = \text{SNR}_{\text{out}} + 10\log(\text{BW}/1\text{Hz}) - 154.01 + \text{NF} - 10\log(\lambda/1\text{micron})$$

where P_s is the launch power, SNR_{out} is the desired signal to noise ratio, BW is the detection bandwidth corresponding to each signal 15 being communicated, NF is the noise figure, and λ is the wavelength of the signal being adjusted. Although this example discusses adjusting launch powers with respect to only one desired signal to noise ratio, signals 15 could, alternatively, be grouped and have launch powers for each group determined with respect to a different signal to noise ratio.

The adjustment to the launched signal power can be accomplished, for example, by adjusting drive current(s) supplying transmitters that generate signals 15. Alternatively, a variable attenuator can be used to

selectively attenuate signals 15 generated at a common launch signal power.

Step 510 may be performed, for example, on an initial system setup, or could be performed throughout operation of system 10 to maintain the SNR despite changes to system characteristics. In one embodiment, step 510 is performed manually. In other embodiments, step 510 can be performed automatically by or with the assistance of link manager 35.

Changes to the launch power spectrum can affect the gain profile of amplifiers in link 25. System 10 adjusts pump powers to at least some amplifiers in link 25 at step 520 to retain a desired gain profile in light of the changes to the launch power spectrum. This can be done, for example, by adjusting drive currents to pump sources, or by adjusting variable attenuators coupled to pump sources.

System 10 determines at step 530 whether variations in the signal to noise ratio for the plurality of signals 15 are within an acceptable tolerance. For example, it may be desired to have the signal to noise ratio for each signal vary by no more than 2.5 decibels. Or, it may be desired to have the signal to noise ratio remain completely flat (e.g., within 0.1 decibels or less) over each of the plurality of signals 15.

If system 10 determines at step 530 that the signal to noise ratio of the signals of interest varies by more than a particular tolerance, system 10 returns to steps 510 and 520, adjusting the launch powers of at least some of signals 15 based at least in part on a noise figure associated with link 25 and adjusting pump powers to retain a desired gain profile. This process continues

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